

SGM41524A/SGM41524B Compact Switch Li+/Polymer Battery Charger with Safe and Reliable Charging

GENERAL DESCRIPTION

The SGM41524A/SGM41524B is compact and efficient Lithium-ion or Lithium-ion polymer (Li+/polymer) battery charger. It can provide power and charge the single-cell battery of a system typically found in compact portable device. An internal switching buck converter regulates the supply input for charging the battery and powering the system even if the battery is absent. The converter can also operate as a simple pass-through switch with no switching if the load and input voltages are close.

A typical application circuit is shown in Figure 2. The SGM41524A/SGM41524B features resistor programmable constant-current and constant-voltage charging capability plus a charge limiting timer and operates in compliance with the BAJ/JEITA safety guide. An NTC (β = 3950K) can be used for battery temperature sensing on top of the internal junction temperature monitoring. The IND status output pin can be connected to LEDs to indicate the operating conditions, such as power input ok (POK), in charging (CHG), VIN over-voltage (POK and CHG alternate blinking) and no power/disabled (OFF). Voltage fold-back on the output is provided to power the system from the input while retaining battery charge and preventing overcharge. Input under-voltage regulation is implemented by reducing the load current such that VIN stays above a minimum when the source is weak. Similarly, the die temperature can be regulated and limited by reducing output power to avoid device or the circuit board being overheated.

These features simplify the system design and ensure safe and reliable operation as well as improved user experience.

The SGM41524A/SGM41524B is delivered in a Green TDFN-2×3-8BL package. The device operates in -40°C to +85°C with +115°C thermal regulation.

FEATURES

- Constant-Current, Constant-Voltage (CC/CV)
 Charging with Floating Time-Out Timer
- Constant-Current Pre-Charge
- Maximum 2.3A Charging for 4.2V to 4.45V Battery
- 1.34MHz Switching Frequency
- Programmable Charge Voltage and Current
- 4.15V Input Voltage Regulation
- Output Voltage Fold-Back Charge Retaining
- Temperature Related Charging Options (NTC Function)
 - SGM41524A: 0°C to 55°C
 - + SGM41524B: 0°C to 45°C
- Typical Peak Efficiency of 92% at 1.5A, V_{VIN} = 5V
- -40°C to +85°C Operating Temperature Range
- Available in a Green TDFN-2×3-8BL Package

APPLICATIONS

Powering and Charge Control of Systems with 500mAh to 6000mAh Li+/Polymer Batteries

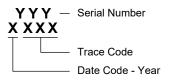


PACKAGE/ORDERING INFORMATION

MODEL	PACKAGE DESCRIPTION	SPECIFIED TEMPERATURE RANGE	ORDERING NUMBER	PACKAGE MARKING	PACKING OPTION
SGM41524A	TDFN-2×3-8BL	-40°C to +85°C	SGM41524AYTDC8G/TR	RD2 XXXX	Tape and Reel, 3000
SGM41524B	TDFN-2×3-8BL	-40°C to +85°C	SGM41524BYTDC8G/TR	RD3 XXXX	Tape and Reel, 3000

MARKING INFORMATION

NOTE: XXXX = Date Code and Trace Code.



Green (RoHS & HSF): SG Micro Corp defines "Green" to mean Pb-Free (RoHS compatible) and free of halogen substances. If you have additional comments or questions, please contact your SGMICRO representative directly.

ABSOLUTE MAXIMUM RATINGS

NOTE:

1. Peak current in IEC61000-4-5 1.2 μ s/50 μ s 2 Ω waveform.

RECOMMENDED OPERATING CONDITIONS

Supply Voltage Range	3.5V to 5.5V
Charge Current Setting Range	0.3A to 2.3A
Operating Junction Temperature Range	40°C to +125°C
Ambient Temperature Range	40°C to +85°C

OVERSTRESS CAUTION

Stresses beyond those listed in Absolute Maximum Ratings may cause permanent damage to the device. Exposure to absolute maximum rating conditions for extended periods may affect reliability. Functional operation of the device at any conditions beyond those indicated in the Recommended Operating Conditions section is not implied.

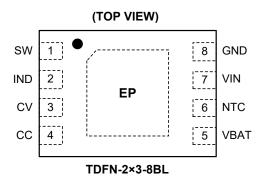
ESD SENSITIVITY CAUTION

This integrated circuit can be damaged if ESD protections are not considered carefully. SGMICRO recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage. ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because even small parametric changes could cause the device not to meet the published specifications.

DISCLAIMER

SG Micro Corp reserves the right to make any change in circuit design, or specifications without prior notice.

PIN CONFIGURATION



PIN DESCRIPTION

PIN	NAME	TYPE	FUNCTION
1	SW	0	Buck Converter Switching Node. Connect to the output inductor.
2	IND	0	Status Indication Output. It can source or sink constant-current when powered (charging or not charging). It can only sink current if no power is applied.
3	CV	I	Charge Voltage Programming Input Pin. Connect a resistor between this pin and ground to select one of the seven charging voltages.
4	CC	I	Charge Current Programming/Charge-Inhibit Input Pin. Connect a resistor between this pin and GND to program the constant charge current I_{CC} , ($R_{CCSET} = K/I_{CC}$). Pull up this pin to a voltage higher than V_{INH} to inhibit and stop charging.
5	VBAT	I	Battery Voltage Sense Input.
6	NTC	ı	NTC Temperature Sensing Input. Connect to an NTC thermistor (β = 3950K) with other end grounded and biased to VIN by a 1.5 × R _{NTC25°C} resistor. Ground this pin if NTC is not used.
7	VIN	Р	Power Input Pin.
8	GND	G	Ground Reference Pin.
Exposed Pad	EP	IC	Exposed Pad. Thermal pad is internally grounded and must be connected to the PCB GND plane.

NOTE:

I = Input, O = Output, G = Ground, P = Power for the Circuit, IC = Internal Connection.

ELECTRICAL CHARACTERISTICS

 $(V_{VIN} = 5V, V_{VBAT} = 3.8V, T_J = -40^{\circ}C$ to +85°C, typical values are at $T_J = +25^{\circ}C$, unless otherwise noted.)

PARAMETERS	SYMBOL	CONDITION	IS	MIN	TYP	MAX	UNITS
O 1/11 D 1 11 T 1 1 1 1	.,	VDAT	T _J = +25°C	5.51	5.67	5.84	.,
Over-Voltage Protection Threshold	V_{OVP}	VBAT open, V _{VIN} = 5V to 6V	$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$	5.49	5.67	5.85	V
M: 1 10 1: 1/1	.,		T _J = +25°C	3.90	4.04	4.17	.,
Minimum Input Operation Voltage	V_{CHGm}	VBAT open, V_{VIN} = 5V to 4V	$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$	3.85	4.04	4.20	V
\(\lambda \) \(\la		IND open, fold-back mode,	T _J = +25°C		15	20	
VIN Supply Current	l _Q	R_{CV} = 1k Ω , set V_{VBAT} = 4.17V, no switching	$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$		15	21	μA
	-	Fold-back mode, $R_{CV} = 1k\Omega$,	T _J = +25°C		0.1	1.4	
Looksons Comment into the VDAT (1)	I _{LKGFLD}	set V_{VBAT} = 4.17V, no switching	$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$		0.1	1.5	μA
Leakage Current into the VBAT ⁽¹⁾		\/\lance=\/\ - 2\/\ta 4.45\/	T _J = +25°C		0.1	1.4	
	I_{LKG}	VIN open, $V_{VBAT} = 3V$ to 4.45V	$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$		0.1	1.5	μA
Charge Loop						•	
Charge Output Regulation Voltage	V	CV nin connected to CND	T _J = +25°C	4.175	4.20	4.225	· V
Charge Output Regulation Voltage	V_{CHG}	CV pin connected to GND	$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$	4.145	4.20	4.255	
Charge Voltage Step	V _{STEP}	$T_J = -40^{\circ}\text{C}$ to $+85^{\circ}\text{C}$			50		mV
The Minimum Voltage Drop between	V_{DROPm}	LDO charge mode, input voltage V _{CHGm} , T _J = -40°C to +85°C	LDO charge mode, input voltage is greater than V _{CHGm} , T _J = -40°C to +85°C		20	40	mV
VIN and VBAT Required for Switch Charging	V_{DROPM}	Switch charge mode, input voltage is greater than V_{CHGm} , $T_J = -40^{\circ}C$ to $+85^{\circ}C$		120	170	230	mV
Charge Voltage Fold-Back when NTC Temperature is out of 10°C to 45°C Range	V_{DEG}	Compare with V _{CHG} in 10°C to 4 temperature range		50		mV	
Charge Current Decrease at NTC	Current Decrease at NTC As percentage		T _J = +25°C		20	1 1	0,1
Temperature Regulation (1)	I_{DEG}	45°C NTC temperature range	$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$		20		%
000 Ti + (1)	DT4		T _J = +25°C	66	68	70	- %
0°C Threshold (1)	DT1	As percentage of V _{VIN}	$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$	65	68	71	
1000 Ti 11 (1)	DT2	As paraentage of \/	T _J = +25°C	56	58	59	- %
10°C Threshold (1)	DIZ	As percentage of V _{VIN}	$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$	55	58	60	
45°C Threshold (1)	DTO		T _J = +25°C	21	23	24	%
45 C Threshold V	DT3	As percentage of V _{VIN}	$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$	20	23	25	70
55°O Throad ald (1)	DT4	As percentage of V _{VIN}	T _J = +25°C	16	17	18	- %
55°C Threshold (1)	DT4		$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$	15	17	19	
Floating Charge Timer Start			T _J = +25°C	96.5	98.0	99.4	0/
Threshold	V_{FLT}	As percentage of V _{CHG}	$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$	96.4	98.0	99.5	%
5.1.5.1.5.1.	ĺ		T _J = +25°C	96.8	97.1	97.5	0/
Fold-Back Retaining Output Voltage	FR	As percentage of V _{CHG}	$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$	96.7	97.1	97.7	- %
	.,		T _J = +25°C	94.0	95.5	97.0	
Recharge Threshold	V_{RR}	As percentage of V _{CHG}	$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$	93.9	95.5	97.1	%
Battery Pre-conditioning Charge		V _{VIN} = 5V,	T _J = +25°C	67	97	128	
Current	I_{PRE}	$V_{VBAT} < 60\% \times V_{CHG}$	$T_{J} = -40^{\circ}\text{C} \text{ to } +85^{\circ}\text{C}$	65	97	130	mA
Battery Pre-conditioning Threshold			T _J = +25°C	57	60	63	0,
Voltage	V_{PRE}	As percentage of V _{CHG}	$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$	56	60	65	%

NOTE: 1. Parameters guaranteed by product characterization.

ELECTRICAL CHARACTERISTICS (continued)

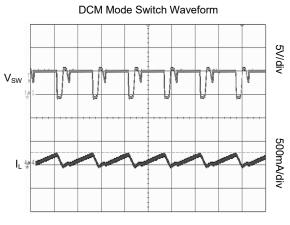
 $(V_{VIN} = 5V, V_{VBAT} = 3.8V, T_J = -40^{\circ}C$ to +85°C, typical values are at $T_J = +25^{\circ}C$, unless otherwise noted.)

PARAMETERS	SYMBOL	CONDITIO	NS	MIN	TYP	MAX	UNITS
Load Pre-Charge Current	I _{LOADPRECHG}	When power-up at V _{VBAT} < 60% × V _{CHG}	$T_{J} = -40^{\circ}\text{C} \text{ to } +85^{\circ}\text{C}$		300		mA
Lood Dro Charge Daried			T _J = +25°C	4.23	5.00	5.77	ma
Load Pre-Charge Period	t _{LOADPRECHG}		$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$	4.05	5.00	5.83	ms
Observed October Dation	17	D 4010 K I D	T _J = +25°C	9450	10000	10500	
Charge Current Setting Ratio	K	$R_{CC} = 10k\Omega, K = I_{CC} \times R_{CCSET}$	$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$	9150	10000	10800	V
Charge Inhibition Voltage Threshold	V_{INH}	Voltage forcing on the CC pin	Voltage forcing on the CC pin to inhibit charging		1.5		V
F 101 0 1		$R_{CC} = 10k\Omega$, $V_{VBAT} = 3.8V$,	T _J = +25°C	0.945	1	1.050	
Fast Charge Current	I _{cc}	$V_{VIN} = 5V$	$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$	0.915	1	1.080	A
Charge Termination Current			T _J = +25°C	95	140	175	
Threshold	I _{RES}		$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$	85	140	190	mA
			T _J = +25°C	77	92	107	
Floating Charge Termination Time	t _{FCOT}		$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$	74	92	108	min
Input Voltage Regulation Threshold	V_{INREG}	V _{VBAT} = 3.8V, V _{VIN} for making charge current to 0	T _J = +25°C	4.00	4.15	4.30	V
			$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$	3.98	4.15	4.32	
Thermal Regulation Threshold (1)	T _{OTR}		1		115		°C
Thermal Shutdown Temperature	T _{SHUT}	Temperature increasing			155		°C
Thermal Shutdown Hysteresis	T _{SHUT_HYST}				20		°C
BAT Voltage Monitoring Period	,		T _J = +25°C	162	192	222	
before Turning into Fold-Back Switch Operation	t _{MON}		$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$	155	192	224	ms
High-side Switch MOSFET	-		T _J = +25°C		140	170	0
On-Resistance between VIN and SW	R _{DS(ON)-H}		$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$		140	200	mΩ
Low-side Switch MOSFET	_		T _J = +25°C		120	150	
On-Resistance between SW and GND	R _{DS(ON)-L}		$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$		120	180	mΩ
Peak Current Limit	I _{PEAK}		$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$		3.2		Α
DW4.0 % 1.1			T _J = +25°C	1.13	1.34	1.55	
PWM Switching Frequency	fs		$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$	1.08	1.34	1.56	MHz
Indication Driving			•				
IND 0: 1 0 (1)			T _J = +25°C	0.7	1.3	1.9	
IND Sink Current (1)	I _{INDSNK}	$V_{VIN} = 5V$	$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$	0.4	1.3	2.5	mA
IND Course Curses (1)		\\ 5\\	T _J = +25°C	0.6	1.3	2.0	A
IND Source Current (1)	INDSRC	V _{VIN} = 5V	$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$	0.4	1.3	2.5	mA
IND DE LED.		1 0 0	T _J = +25°C	162	192	222	
IND Blink Period	t _{BLINK}	Input OVP state	$T_J = -40^{\circ}C \text{ to } +85^{\circ}C$	155	192	224	ms

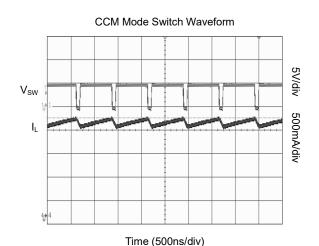
NOTE: 1. Parameters guaranteed by product characterization.

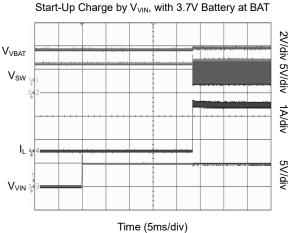
TYPICAL PERFORMANCE CHARACTERISTICS

 V_{VIN} = 5V, V_{VBAT} = 3.8V, T_J = +25°C, unless otherwise noted.

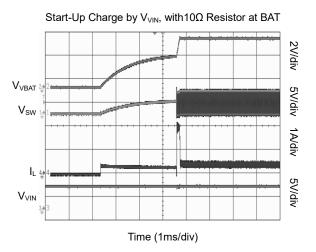




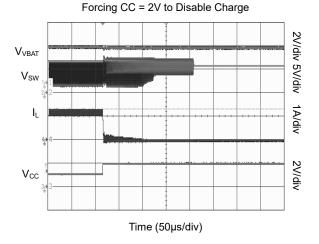


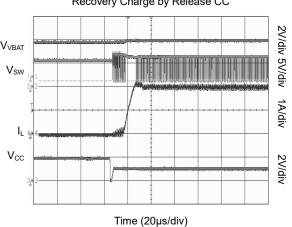


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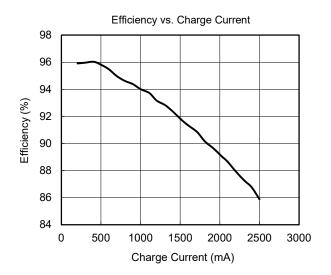
Recovery Charge by Release CC

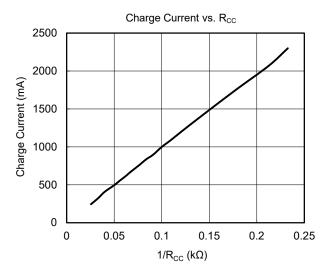




TYPICAL PERFORMANCE CHARACTERISTICS (continued)

 V_{VIN} = 5V, V_{VBAT} = 3.8V, T_J = +25°C, unless otherwise noted.





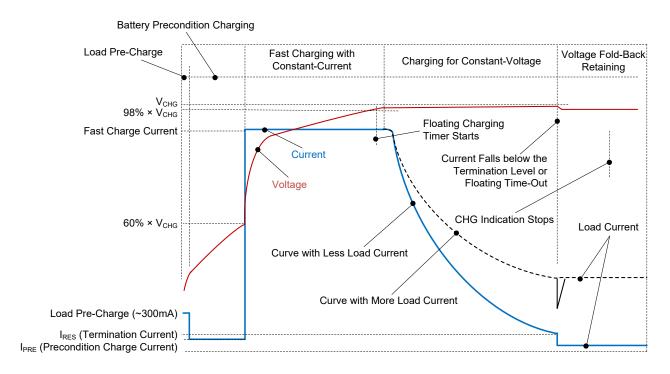


Figure 1. Charging Voltage/Current Profile

TYPICAL APPLICATION

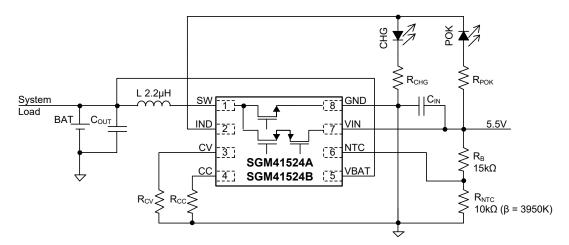


Figure 2. Typical Application Circuit

FUNCTIONAL BLOCK DIAGRAM

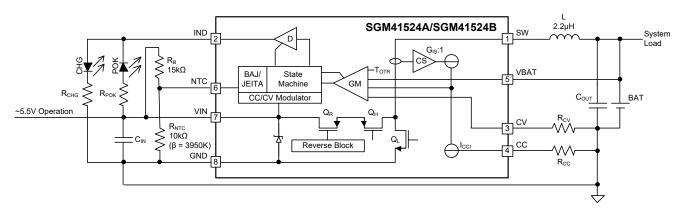


Figure 3. Block Diagram

ESSENTIAL SEQUENCE

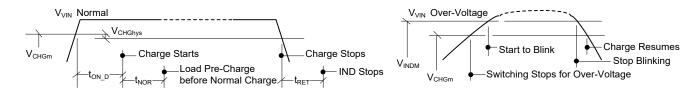


Figure 4. Essential On/Off Timing

FUNCTION DESCRIPTION AND APPLICATION

The SGM41524A/SGM41524B is a general purpose stand-alone switch mode charger device designed for powering systems using Li+/polymer rechargeable batteries. Several features are provided including charge voltage and current programming and status indication. Input voltage and die temperature are constantly monitored to prevent output power failure. If the input supply voltage drops too low, the device reduces the output power to reduce loading on the input and prevent further drop and power failure. Similarly, if the junction is overheated by heavy load, the output power is reduced to prevent thermal shutdown and system power failure. It is also capable for various charging modes like constant-current, constant-voltage, constant-current precharging, and trickle charging (when input source is weak).

This device does not have a separate battery switch to connect or disconnect the battery from the system (load). However, it uses voltage fold-back retaining for battery safety and lifetime extension when the battery is fully charged and input power is present. With this method, battery energy loss is lower because there is no switch in the discharge path. The only disadvantage is that if the battery voltage is excessively low, then start the system instantly is not possible because charge path cannot be separated and it may take a few minutes to charge the battery and reach to an adequate voltage level to start the load system.

Power-Up with Low/No Battery

If the battery is not attached or its voltage is less than 60% of V_{CHG} ($V_{VBAT} < 0.6 V_{CHG}$), the device feeds the VBAT with a current limited to less than 300mA for about 5ms to pre-charge the battery and system load before it goes into battery pre-conditioning charging state. This pre-charge period can increase the voltage of a 500 μ F capacitor (between VBAT and GND) for up to 3V before the device starts to deliver the lower pre-conditioning charge current (97mA TYP).

If the load is started before fast charge phase, the supply capacity will be limited to the pre-conditioning charge current for a relatively long time. The initial 5ms pre-charge period can quickly bring the device to the fast charge or even fold-back phase when there is no battery attached and provide enough power for the system operation in a short time.

Charging Profile and Fold-Back Retaining

The charging profile is shown in Figure 1. When the battery voltage is less than $0.6V_{CHG}$, the output current is regulated to a low and safe pre-conditioning level (I_{PRE}). On the other hand

when the input supply is present and charge is complete, the output goes to the safe voltage fold-back mode for powering the load. In this mode, the output current is limited to less than peak current limit (I_{PEAK}) and not to the programmed charge current limit. Figure 5 shows the load transient response of the evaluation board circuit whose charge current is programmed for less than the load current.

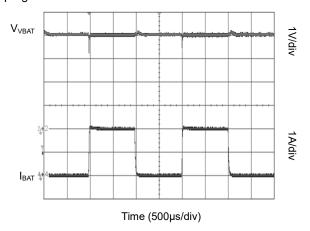


Figure 5. Load Transient Response in Fold-Back Supply

The input voltage is monitored during charging. If the source is weak and cannot maintain its voltage under heavy load, the charging current is reduced to avoid system power collapse due to input voltage drop.

The charge is considered full if the battery voltage exceeds the floating charge timer start threshold (V_{FLT}) and the charge current drops below the charge termination current (I_{RES}) or if the floating charge timer runs out of time (I_{FCOT}).

When the full charge status is detected, the output will drop to the fold-back voltage specified by fold-back retaining output voltage ratio (FR ratio is typically 97.1% of V_{CHG} as specified in the EC table) and converter continues to work but indicator shows "not charging". If the battery voltage is higher than fold-back level, the switching will stop. VBAT is monitored periodically and if it drops below that level the buck converter starts to operate and regulate the output to the fold-back level. The full charge state is continued until the input power is recycled or if the battery voltage drops below the recharge threshold (V_{RR}).

If the voltage drop between the input and output (V_{DROP}) is small and less than V_{DROPm} , the device goes into forward diode state and stops switching. Switching is resumed if the V_{DROP} exceeds V_{DROPM} level. The CHG indication will turn off if the time of $V_{DROP} < V_{DROPm}$ is longer than the retaining time (t_{RET}).

FUNCTION DESCRIPTION AND APPLICATION (continued)

Charge Current Programming and Turn-Off

Charge current is programmed by R_{CCSET} resistance by $R_{\text{CCSET}} = K/I_{\text{CC}}$ where the K is charge current setting ratio which is typically 10000V as specified in the EC table.

Pulling the CC pin to a voltage level higher than V_{INH} turns the device off (disabled). When this pin is released the device resumes the status before being inhibited.

Charge Voltage Programming

Charging voltage can be programmed in one of the 7 preset values by setting a voltage on CV pin. A 50μ A current source is internally connected to CV pin. Programming can be done by directly applying a voltage to the CV pin, or by connecting a resistor to GND that results in the same voltage as shown in Table 1.

Table 1. Conditions for Selecting a Charging Voltage

Charging Voltage (V)	Forcing Voltage (V)	Separation Thresholds (V)	Grounding Resistance (kΩ)
4.2	GND	< 0.4 or > 2.4	Short or Open
4.25	0.6	0.4 to 0.8	12
4.3	1.0	0.8 to 1.2	20
4.35	1.4	1.2 to 1.6	28
4.4	1.8	1.6 to 2.0	36
4.45	2.2	2.0 to 2.4	44

NOTE: Sourcing current out of the CV is 50µA typically.

BAJ/JEITA Charging Extending and Safety

This device implements the battery temperature related charging control in compliance with the BAJ/JEITA guide on safe use of secondary Lithium-ion batteries. An NTC (β = 3950K) can be used as shown in Figure 2 (or Figure 3) for battery temperature sensing.

As specified in Table 2 and Table 3, the charging voltage and current are reduced when the sensed battery temperature is out of the preferred charging range (10°C to 45°C). When the temperature is too high (above 55°C for SGM41524A and above 45°C for SGM41524B) or too low (less than 0°C), the device stops charging.

Table 2. SGM41524A Temperature Related Charging Control

Temperature Range	Charging Voltage	Charging Current
Low range, < 0°C.		0
Low charging range, 0°C to 10°C	V _{CHG} - 50mV	20% I _{CC}
Recommended charging range, 10°C to 45°C	V_{CHG}	I _{cc}
High charging range, 45°C to 55°C	V _{CHG} - 50mV	20% I _{CC}
High range, > 55°C	_	0

Table 3. SGM41524B Temperature Related Charging Control

Temperature Range	Charging Voltage	Charging Current
Low range, < 0°C.	_	0
Low charging range, 0°C to 10°C	V _{CHG} - 50mV	20% I _{CC}
Recommended charging range, 10°C to 45°C	V_{CHG}	Icc
High range, > 45°C	_	0

NOTE: The V_{CHG} and the I_{CC} (charging voltage and current) are selected in accordance with the battery's specification.

If NTC feature is not used, connect the NTC pin to ground. The device checks for grounded NTC pin once during the start-up when the input voltage is exceeding 2.7V.

Indication and Status Reading

The IND output can have 4 states to show different conditions: (1) Low (sink current) to indicate the input power is available (or not charging); (2) High (source current) to indicate the device is in charging; (3) Hi-Z (open) for indicating no power is available (when $V_{VIN} < V_{CHGm}$) or when it is turned off by pulling the CC pin voltage up; and (4) blinking or alternatingly Low and High (sinking and sourcing current) if an input over-voltage occurs. IND voltage can be used as a signal for the host or other circuit for status detection.

Note that in the high impedance state (Hi-Z), the POK LED and CHG LED are forward biased by the input voltage (all in series) and they can turn on depending on the drive current determined by the LED forward voltages and series resistances.

FUNCTION DESCRIPTION AND APPLICATION (continued)

Input Voltage Regulation and Thermal Regulation

To prevent power shutdown, the output current is gradually reduced if VIN drops close to the minimum (V_{CHGm}). Output current eventually reaches to zero when VIN falls to V_{CHGm} level. Similarly, if the junction temperature increases close to its maximum (T_{OTR}), the output current is progressively reduced and will reach to zero when the temperature reaches to T_{OTR} .

Component Selection

Inductor Selection

Small inductors and capacitors can be chosen thanks to the high operating switching frequency of the 1.34MHz. Select an inductor with a saturation current a little bit higher than the charging current (I_{CHG}) plus half the ripple current peak to peak magnitude (I_{RIPPLE}):

$$I_{SAT} \ge I_{CHG} + (1/2) I_{RIPPLE}$$
 (1)

The inductor ripple current depends on the input voltage (V_{VBUS}) , the duty cycle $(D = V_{VBAT}/V_{VBUS})$, the switching frequency (f_S) and the inductance (L). In CCM (e.g. full load):

$$I_{RIPPLE} = \frac{V_{VBUS} \times D \times (1 - D)}{f_{s} \times L}$$
 (2)

The maximum inductor ripple current occurs when the duty cycle (D) is 0.5 or near. Typically, the inductor ripple is designed in the range between 20% and 40% of the maximum charging current as a trade-off between inductor size and efficiency. Smaller inductor results in higher ripple (AC) current flowing into the capacitor and switches and can reduce efficiency.

Input Capacitor

Choose the input capacitance with enough RMS current rating to decouple input switching AC currents away from input. Low ESR ceramic capacitor such as X5R or X7R is preferred for input decoupling. Typically, $10\mu\text{F}$ capacitance is suitable for 1A to 2A charging current. Keep the capacitor(s) close to VIN and GND pins to minimize the parasitic inductance in the input ripple current circulation path. In the worst-case, the RMS of the ripple current is half of the DC charging current (I_{CHG}) when duty cycle is D = 50%. If the converter does not operate at 50% duty cycle, then the worst-case occurs when duty cycle is closest to 50%. The input RMS current (I_{CIN}) can be estimated by Equation 3.

$$I_{CIN} = I_{CHG} \times \sqrt{D \times (1 - D)}$$
 (3)

Output Capacitor

A few factors must be considered to design the output capacitance. First, the SGM41524A/B has the internal loop compensation for the buck converter that is optimized for ceramic output capacitance larger than $10\mu F$. The output capacitor (C_{OUT}) circulates the output ripple current and prevents it from going into the battery. Having AC current in the battery results in extra heating and lower lifetime.

Equation 4 gives the output capacitor RMS current I_{COUT} when no battery is attached.

$$I_{\text{COUT}} = \frac{I_{\text{RIPPLE}}}{2 \times \sqrt{3}} \approx 0.29 \times I_{\text{RIPPLE}}$$
 (4)

The RMS ripple voltage in worst case is calculated as:

$$V_{RIPPLE} = \frac{I_{RIPPLE}}{2 \times \pi \times f_{S} \times C_{OUT}}$$
 (5)

The capacitance should be selected large enough for meeting the system requirement for acceptable V_{RPPLE} .

In the system design, operation with no battery must be considered carefully. Typically, the presence of the battery helps in filtering of the sags and ripples and provides peak energy demands when load surges occur. When the battery is absent, a relatively large capacitor is needed to have proper performance.

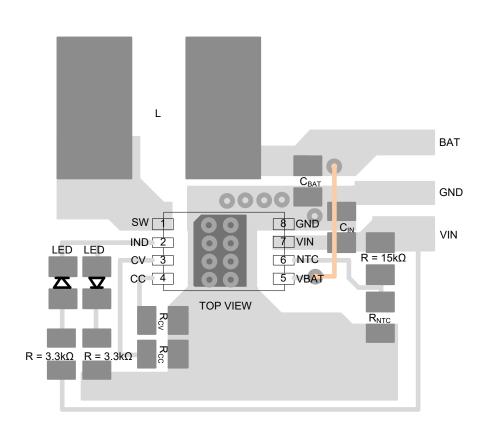
Besides the V_{RIPPLE} requirement, the load starting inrush current is another factor to consider for output capacitor selection. If at the beginning the device turns into fold-back, the converter does not start switching as the output capacitor holds the voltage higher than the fold-back retaining voltage. VBAT voltage is monitored periodically and as long as it is above fold-back voltage, it is only the output capacitor that powers the system in the absence of the battery. The capacitance should be large enough to maintain the VBAT voltage and prevent dropping below minimum system requirement before the switching fold-back mode supply operation starts. The capacitance for fulfilling this requirement is highly dependent on how the load starts, including its timing, start current and acceptable voltage drop. Verification with a prototype is recommended if operation without a battery is considered.

FUNCTION DESCRIPTION AND APPLICATION (continued)

Layout Guide

- 1. Place VIN capacitor close to the VIN pin and GND pin.
- 2. Place the inductor terminal close to the SW pin and minimize the copper area of switching node trace. Do not use multiple layers for this connection.
- 3. Minimize the return loop area and ripple current path length through the inductor and the output capacitor(s) to the device GND pin.
- 4. Use copper plane for power GND and place multiple via between top and bottom GND plane for better heat dissipation and noise immunity.

PCB Layout Example



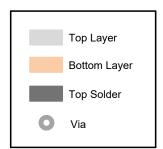


Figure 6. Typical PCB Layout

FUNCTION DESCRIPTION AND APPLICATION (continued)

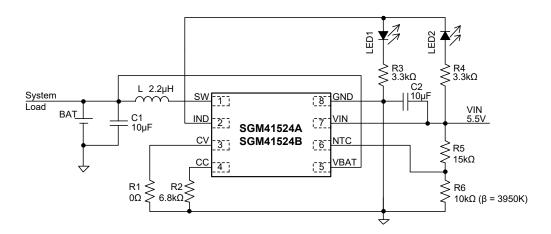


Figure 7. Typical Application Circuit, Programmed for Charge Current of Icc = 1.47A, and Voltage of V_{CHG} = 4.20V

Table 4. Bill of Materials for Typical Application Circuit

Designator	Quantity	Description	Size	Maker	Part Number
U1	1	Switch Li+/Polymer Battery Charger	TDFN-2×3-8BL	SGMICRO	SGM41524A/SGM41524B
L1	1	Ind, 2.2μH, Irms = 4.3A, Isat = 6.1A, DCR = 40mΩ	4.0*4.0*2.0mm	Sunlord	WPN4020H2R2MT
C1, C2	2	Cap, Cerm, 10µF, 10V, X5R	0603	SAMSUNG	
R1	1	Res, 0Ω, 1%	0603	UniOhm	
R2	1	Res, 6.8kΩ, 1%	0603	UniOhm	
R3, R4	2	Res, 3.3kΩ, 5%	0603	UniOhm	
R5	1	Res, 15kΩ, 1%	0603	UniOhm	
R6	1	NTC, 10kΩ, 1%, β = 3950K	0603	Sunlord	SDNT1608X103F3950FTF
LED1, LED2	2	Chip Light Emitting Diode, Blue	0603	Nationstar	FC-DA1608BK-470H10

REVISION HISTORY

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes non Original (MARCH 2021) to REV	anges from Original (MARCH 2021) to REV	.м
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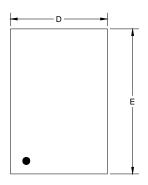
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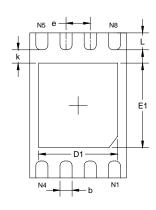
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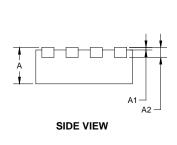
PACKAGE OUTLINE DIMENSIONS TDFN-2×3-8BL

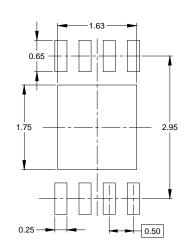


TOP VIEW



BOTTOM VIEW



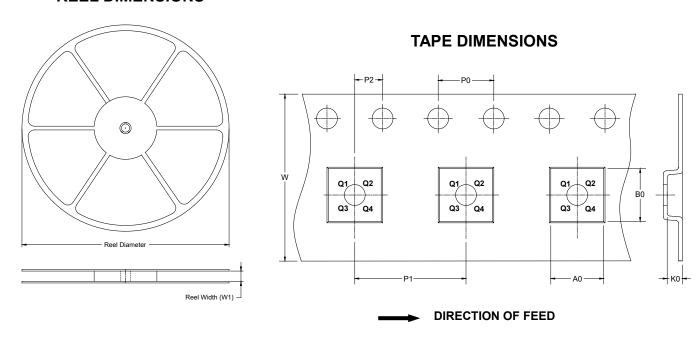


RECOMMENDED LAND PATTERN (Unit: mm)

Symbol	Dimensions In Millimeters		Dimensions In Inches		
	MIN	MAX	MIN	MAX	
А	0.700	0.800	0.028	0.031	
A1	0.000	0.050	0.000	0.002	
A2	0.203	REF	0.008 REF		
D	1.950	2.050	0.077	0.081	
D1	1.530	1.730	0.060	0.068	
E	2.950	3.050	0.116	0.120	
E1	1.650	1.850	0.065	0.073	
b	0.200	0.300	0.008	0.012	
е	0.500 BSC 0.020 BSC		BSC		
k	0.250 REF 0.010 REF		REF		
L	0.300	0.450	0.012	0.018	

TAPE AND REEL INFORMATION

REEL DIMENSIONS

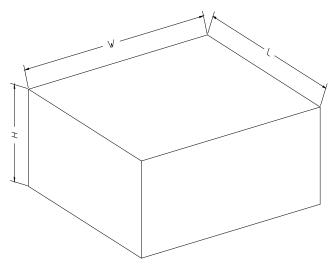


NOTE: The picture is only for reference. Please make the object as the standard.

KEY PARAMETER LIST OF TAPE AND REEL

Package Type	Reel Diameter	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P0 (mm)	P1 (mm)	P2 (mm)	W (mm)	Pin1 Quadrant
TDFN-2×3-8BL	7"	9.5	2.30	3.30	1.10	4.0	4.0	2.0	8.0	Q2

CARTON BOX DIMENSIONS



NOTE: The picture is only for reference. Please make the object as the standard.

KEY PARAMETER LIST OF CARTON BOX

Reel Type	Length (mm)	Width (mm)	Height (mm)	Pizza/Carton	
7" (Option)	368	227	224	8	
7"	442	410	224	18	